

Application Note

Low loss Series Regulator IC

SI-3000KF series

Not Recommended for New Designs

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SANKEN ELECTRIC CO., LTD.

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1. General Description

The SI-3000KF is a series regulator IC using a hyposaturation type PNP bipolar transistor in the power section and it can be used with the low difference of input/output voltages. It is provided with an ON / OFF terminal which operates in Active High mode and the current consumption of circuits at OFF time is zero.

● 1-1 Features

- Output current 1A
Output current is 1A at maximum with the outline of TO220F-5.
- Hyposaturation ($V_{dif} = 0.6 V_{max} / I_o = 1A$)
It can be designed with low difference of input/output voltages.
- ON/OFF function
The ON/OFF terminal which can be directly controlled by TTL logic signals is provided.
- Low current consumption
Current consumption of circuits at OFF time is zero.
Dark current at no load is $600\mu A$ at maximum.
- High ripple attenuation ratio
 $75dB: F = 100 - 120kHz$ at $V_o = 5V$
- Built-in Overcurrent protection / Thermal shutdown
The automatic restoration and Foldback type overcurrent protection and Thermal shutdown circuit are built in.

● 1-2 Application

For on-board local power supplies, power supplies for OA equipment, stabilization of secondary output voltage of regulator and power supply for communication equipment

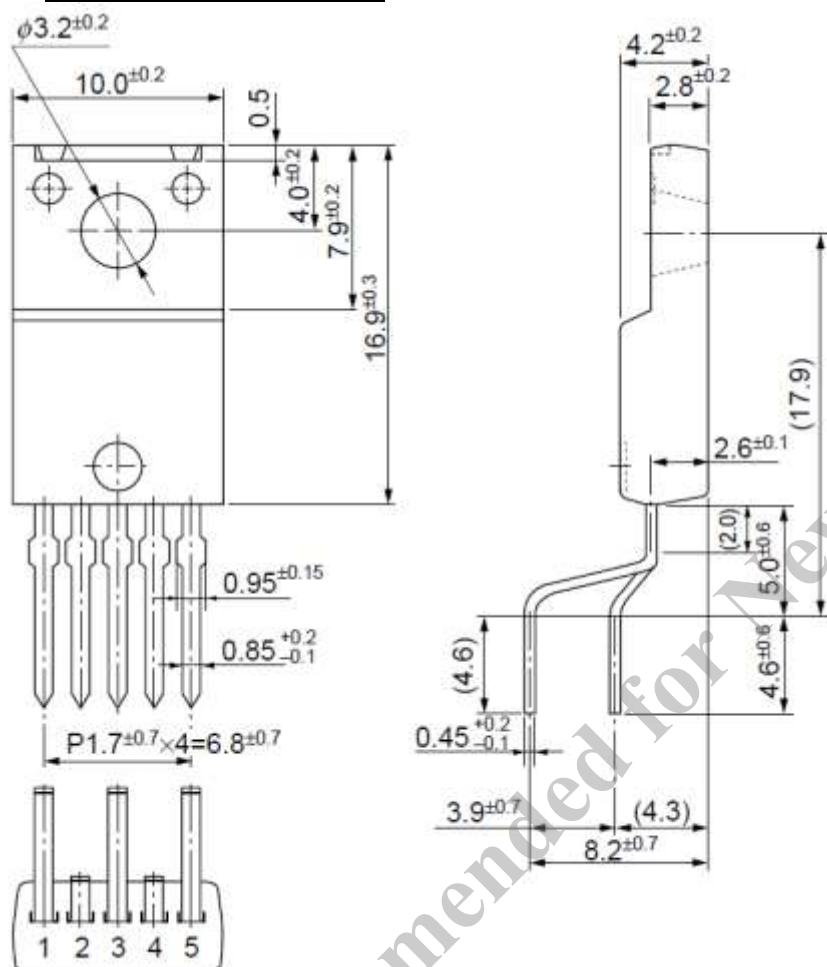
● 1-3 Type

- Type: Semiconductor integrated circuits (monolithic IC)
- Structure: Resin molding type (transfer molding)

2. Specification

Unit: mm

● 2-1 Package Information



Pin assignment

1. V_c
2. V_{IN}
3. GND
4. V_{out}
5. ADJ

Resin sealed type

Non-combustibility: UL standards 94V-0

Product mass: about 2.3 g

● 2-2 Ratings

2-2-1 Absolute Maximum Ratings

Ta = 25°C

Parameter	Symbol	Ratings	Unit	Remarks
		SI-3010KF		
DC Input Voltage	V _{IN}	35*1	V	
Output Control Terminal Voltage	V _c	V _{IN}	V	
DC Output Current	I _o	1.0	A	
Power Dissipation	P _{D1}	16.6	W	With infinite heatsink
	P _{D2}	1.72	W	Without heatsink, stand-alone operation
Junction Temperature	T _J	-40 to +125	°C	
Storage Temperature	T _{stg}	-40 to +125	°C	
Operating Ambient Temperature	T _{op}	-40 to +100	°C	
Thermal Resistance (Junction to Case)	θ _{j-c}	6.0	°C/W	
Thermal Resistance (Junction to Ambient Air)	θ _{j-a}	58	°C/W	Without heatsink, stand-alone operation

*1: A built-in input-overvoltage-protection circuit shuts down the output voltage at the Input Overvoltage Shutdown Voltage of the electrical characteristics.

2-2-2 Recommended Conditions

Parameter	Symbol	Ratings	Unit
		SI-3010KF	
Input Voltage Range	V _{IN}	2.4*2 to 27*1	V
Output Current Range	I _o	0 to 1.0*3	A
Output Voltage Variable Range	V _{oADU}	1.1 to 1.8	V
Operating Ambient Temperature	T _{op}	-30 to +85	°C
Operating Junction Temperature	T _J	-20 to +100	°C

*1: V_{IN} (max) and I_o (max) are restricted by the relationship P_D (max) = (V_{IN} - V_o) × I_o = 16.6W.

*2: Refer to the Dropout Voltage parameter.

2-2-3 Electrical Characteristics

Ta = 25°C

Parameter	Symbol	Ratings			Unit	
		SI-3010KF				
		min.	typ.	max.		
Reference Voltage	V _{ADJ}	0.98	1.00	1.02	V	
	Conditions	V _{IN} =7V, I _O =0.01A, V _C =2V, V _O =5A				
Line Regulation	ΔV _{OLINE}			30	mV	
	Conditions	V _{IN} =6 to 15V, I _O =0.01A, V _C =2V, V _O =5A				
Load Regulation	ΔV _{OLOAD}			75	mV	
	Conditions	V _{IN} =7V, I _O =0 to 1A, V _C =2V, V _O =5A				
Dropout Voltage	V _{DF}			0.3	V	
	Conditions	I _O =0.5A, V _C =2V, V _O =5V				
	Conditions	I _O =1.0A, V _C =2V, V _O =5V				
Quiescent Circuit Current	I _q			600	μA	
	Conditions	V _{IN} =7V, I _O =0A, V _C =2V				
Circuit Current at Output OFF	I _q (OFF)			1	μA	
	Conditions	V _{IN} =7V, V _C =0V				
Temperature Coefficient of Output Voltage	ΔV _O /ΔT _a			±0.5	mV/°C	
	Conditions	V _{IN} =7V, I _O =0.01A, V _C =2V, T _J =0 to 100°C, V _O =2.5V				
Ripple Rejection	RR _{rej}		75		dB	
	Conditions	V _{IN} =7V, I _O =0.1A, V _C =2V, f=100 to 120Hz, V _O =5V				
Overcurrent Protection Starting Current ^{*3}	I _{S1}	1.1			A	
	Conditions	V _{IN} =7V, V _C =2V				
V _C Terminal	Control Voltage (Output ON) ^{*4}	V _C , I _H	2		V	
		Conditions	V _{IN} =7V			
	Control Voltage (Output OFF)	V _C , I _L			0.8	V
		Conditions	V _{IN} =7V			
	Control Current (Output ON)	I _C , I _H			40	μA
		Conditions	V _{IN} =7V, V _C =2V			
Control Current (Output OFF)	I _C , I _L	-5	0		μA	
	Conditions	V _{IN} =7V, V _C =0V				
Input Overvoltage Shutdown Voltage	V _{OV}	33			V	
	Conditions	I _O =0.01A				

*1: I_{S1} is specified at the 5% drop point of output voltage V_O on the condition that V_{IN} = overcurrent protection starting current, I_O = 10mA.

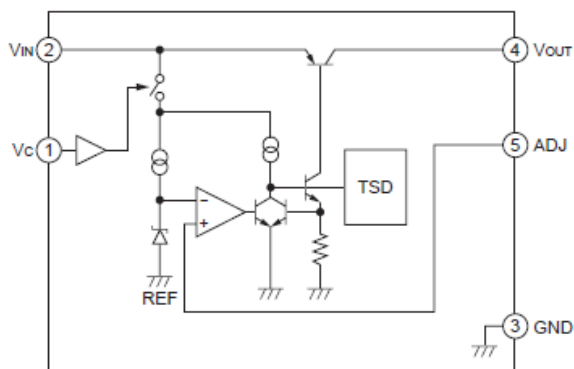
*2: Output is OFF when the output control terminal V_C is open. Each input level is equivalent to LS-TTL level. Therefore, the device can be driven directly by LS-TTLs.

*3: SI-3000KFE cannot be used in the following applications because the built-in foldback-type overcurrent protection may cause errors during start-up stage;

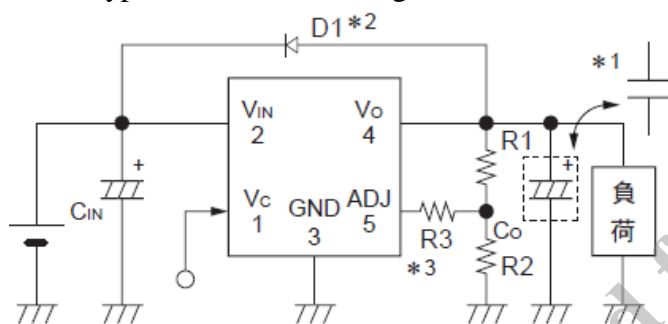
(1) Constant current load (2) Positive and negative power supply (3) Series-connected power supply (4) V_O adjustment by raising ground voltage

● 2-3 Circuit Diagram

2-3-1 Block Diagram



2-3-2 Typical Connection Diagram



C_{IN} : Input Capacitor (over $22\mu\text{F}$)

C_o : Output Capacitor (over $47\mu\text{F}$)

In the case that capacitors having low ESR such as ceramic capacitors are used for output capacitors, they may oscillate.

*1: D1: Reverse biased protection diodes

In the case of reverse bias between input and output, this diode will be required.

(Recommended diodes: SJPL-H2 made by Sanken)

R1, R2: resistors for setting output voltages

Output voltages can be adjusted by connecting R1 and R2 as shown in the above figure.

R2: $10\text{k}\Omega$ is recommended.

$$R1 = (V_o - V_{ADJ}) / (V_{ADJ} / R2)$$

*2: In the case that $V_o \leq 1.5\text{V}$ is set, R3 should be inserted. $10\text{k}\Omega$ is recommended for R3.

3. Operational Description

● 3-1 Voltage Control

In the SI-3000KF series, the driving circuit is controlled by comparing the reference voltage with the ADJ terminal voltage (voltage divided by V_o detection resistor in fixed output products) to stabilize the output voltage by varying the voltage between the emitter and collector of a main PNP power transistor. The product of voltage between emitter and collector and the output current at this moment is consumed as heat.

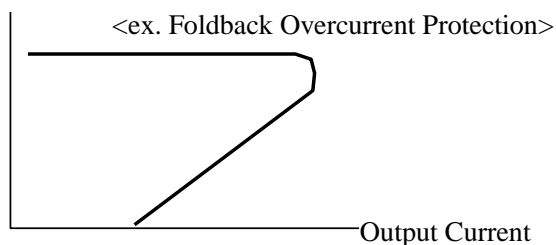
● 3-2 Overcurrent Protection

3-2-1 Overcurrent Protection Characterization for SI-3010KF

The foldback type overcurrent protection function is provided in the SI-3010KF. After operation of the overcurrent protection function, if the load resistance decreases and the output voltage drops, the output current of products is squeezed to reduce the increase of loss. However, in the case of the foldback type overcurrent protection function, since current limiting is also made at start-up, the function may not be used for the following applications, as it may cause a start-up error.

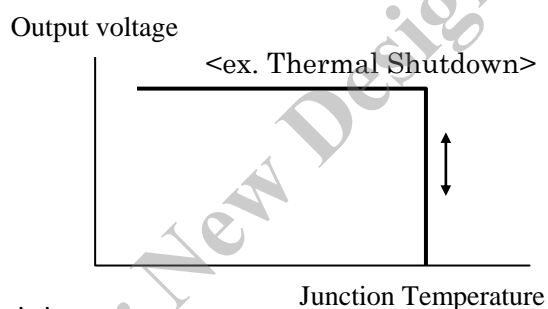
- (1) Constant current loads
- (2) Plus/minus power supply
- (3) DC power supply
- (4) Output voltage adjustment by grounding-up

Output voltage



● 3-3 Thermal Shutdown

This IC is provided with the overheat protection circuit which detects the semiconductor junction temperature of the IC to limit the driving current, when the junction temperature exceeds the set value (around 150°C). Since the minimum operating temperature of the overheat protection circuit is 130°C, the thermal design of $T_j < 125^\circ\text{C}$ is required. Since the overheat protection has no hysteresis, as soon as the overload state is released and T_j falls below the set temperature, the normal operation is automatically restored. When the overheat protection function is operated in the overload state, the output voltage falls, but at the same time the output current is decreased and in the consequence, overheat protection operation and automatic restoration are repeated in a short interval, resulting eventually in the waveforms of output voltage oscillation.



*Note for thermal shutdown characteristic

This circuit protects the IC against overheat resulting from the instantaneous short circuit, but it should be noted that this function does not assure the operation including reliability in the state that overheat continues due to long time short circuit.

Not Recommended for New Designs

4. Cautions

● 4-1 External Components

4-1-1 Input Capacitor C_{IN}

The input capacitor is required to eliminate noise and stabilize the operation and values of 0.47μF - 22μF are recommended. Any of ceramic capacitors or electrolytic ones may be used for the input capacitor.

4-1-2 Output Capacitor C_O

In the output capacitor C_O, larger capacitance than the recommended value is required for phase compensation. Equivalent series resistance values (ESR) of capacitors are limited, and depending on products, therefore the type of recommended capacitors is limited.

- Recommended values of SI-3010KF: $2\Omega > \text{ESR} > 0.2\Omega$

It is recommended to use electrolytic capacitors. When capacitors with extremely high ESR such as ceramic capacitors, functional polymer capacitors etc., are used, phase margin is decreased, possibly causing the oscillation of output voltage. Therefore these capacitors can not be used.

4-1-3 Reverse bias protection diode D1

In the case of falling-down of the input voltage, it is recommended to insert a protection diode D1 against the reverse bias between input and output. However, in the case of setting the V_{out} < 3.3V or lower, D1 is not required including the case of reverse bias. In order to select a suitable D1, it should be taken into consideration that the diode has adequate forward current withstand voltage against the instantaneous discharge of energy stored in C_{out}.

The permissible value of the forward current per unit time of diode is specified in I_{FSM} (A) and in the case of our diode, it is specified at 50Hz half wave (10ms), but it should be noted that different companies may specify different times. The selection of diode should be made by converting the specified time into the actual discharging time so as to meet the required I_{FSM} (A). The discharging time of C_O is normally shorter than 1ms, but it is recommended to do the conversion with 1ms in consideration of margin.

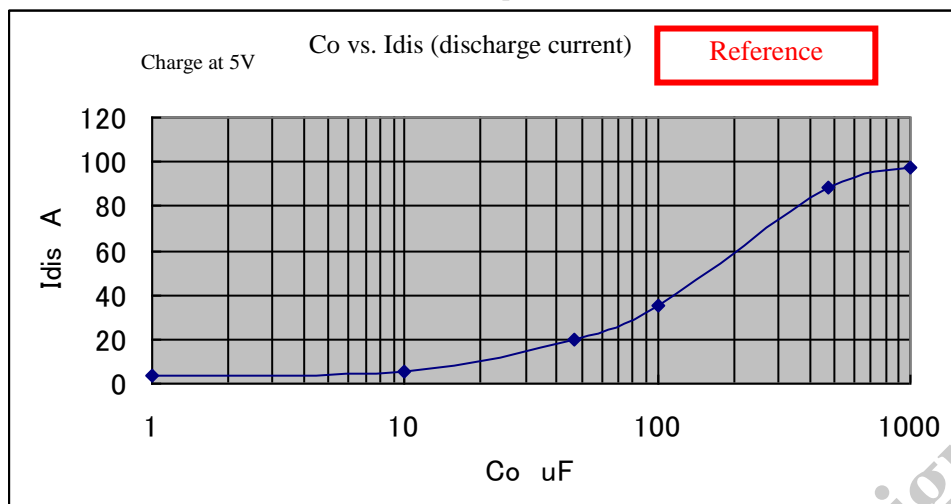
For conversion into I_{FSM}, calculation should be made by using the equations (1) and (2).

$$\left(\frac{I_{FSM}}{\sqrt{2}}\right)^2 * t1 = X \quad \text{--- (1) As for } I_{FSM}, \text{ please refer to the catalog of each company.}$$

t1 = specified time in catalog of each company

$$\text{Converted IFSM} = \sqrt{\frac{2 * X}{t2}} \quad \text{--- (2) } t2: \text{ converted time (discharging time of } C_O)$$

<Graph 1>



On the assumption of $C_{out} = 470\mu\text{F}$, I_{FSM} of around 90A or more (in 1ms time period) is required and according to our specifications of diode, I_{FSM} is specified for 10ms, therefore the diode of 30A has the tolerated dose of 94.8A (in 1ms) to prove that it is usable.

● 4-2 Pattern Design Notes

4-2-1 Input / Output Capacitor

The input capacitor C1 and the output capacitor C2 should be connected to the IC as close as possible. If the rectifying capacitor for AC rectifier circuit is on the input side, it can be used as an input capacitor. However, if it is not close to the IC, the input capacitor should be connected in addition to the rectifying capacitor.

4-2-2 ADJ Terminal (Output Voltage Set-up for SI-3010KF)

The ADJ terminal is a feedback detection terminal for controlling the output voltage. The output voltage set-up is achieved by connecting R1 and R2.

SI-3010KF: it should be set in a manner that I_{ADJ} is around $100\mu\text{A}$.

R1, R2 and output voltage can be obtained by the following equations:

$$I_{ADJ} = V_{ADJ} / R2 \quad \left[*V_{ADJ} = 1.0\text{V} \pm 2\% \text{ (SI-3010KF), } R2 = 10\text{k}\Omega \text{ recommended} \right]$$

$$R1 = (V_o - V_{ADJ}) / I_{ADJ} \quad R2 = V_{ADJ} / I_{ADJ}$$

$$V_{out} = R1 \times (V_{ADJ} / R2) + V_{ADJ}$$

5. Applications

● 5-1 Output ON / OFF Control

The ON/OFF control of output can be made by directly applying voltage to No. 1 Vc terminal. When the Vc terminal is open, the operation is in OFF. The Vc terminal is in OFF below 0.8V and in ON at above 2V.

● 5-2 Thermal Design

5-2-1 Calculation of heat dissipation

As the junction temperature Tj (MAX) is a product-specific value, it must be observed strictly. For this purpose, heat sink design (decision of θ_{fin}) which is appropriate for Pd (MAX) and Ta MAX is required. This is graphically shown in the heat derating curve for easy understanding. The heat dissipation design is done in the following procedure.

- 1) The highest ambient temperature in the set Ta MAX is obtained.
- 2) The maximum loss PdMAX which varies the input/output conditions is obtained.

$$Pd = (V_{IN} - V_{OUT}) \times I_{OUT}$$

- 3) The area of heat sink is determined from the intersection point in the heat derating curve.

The required thermal resistance of the heat sink can be also calculated. The thermal resistance required for the heat sink is obtained by the following equation:

$$\theta_{fin} = \frac{T_j - T_a}{Pd} - \theta_{j-c}$$

An example of heat calculation for using SI-3010KF under the conditions of $V_{IN} = 8V$, $V_o = 5V$, $I_o = 1A$ and $T_a = 85^\circ C$ is shown below. As $T_{jmax} = 125^\circ C$,

$$Pd = (8V - 5V) \times 1A = 3W \quad \theta_{j-c}: 6^\circ C/W$$

$$\theta_{fin} = \frac{125 - 85}{3} - 6 \doteq 7.33^\circ C/W$$

As a result, the heat sink with the thermal resistance of $7.33^\circ C/W$ or less is required.

As described above, the heat sink is determined, but the derating of 10 - 20% or more is used. Actually, heat dissipation effect significantly changes depending on the difference in component mounting. Therefore, heat sink temperature or case temperature should be checked with the heat sink mounted.

5-2-2 Installation to Heat Sink

Selection of silicon grease

When the SI-8000FFE is installed to the heat sink, silicon grease should be thinly and evenly coated between the IC and heat sink. Without coating, thermal resistance θ_i is significantly increased because of

contact failure due to micro concavity/convexity between the backside of the IC and the surface of the heat sink to accelerate the heating of the IC, resulting in shorter life of the IC.

In some kind of silicon grease to be used, oil component may be separated to penetrate into the IC, resulting in the deformation of packages or the adverse effect on built-in elements. Any other silicon grease than one based on the modified silicon oil shall not be used. The recommended silicon greases are as follows:

Sanken's recommended silicon greases:

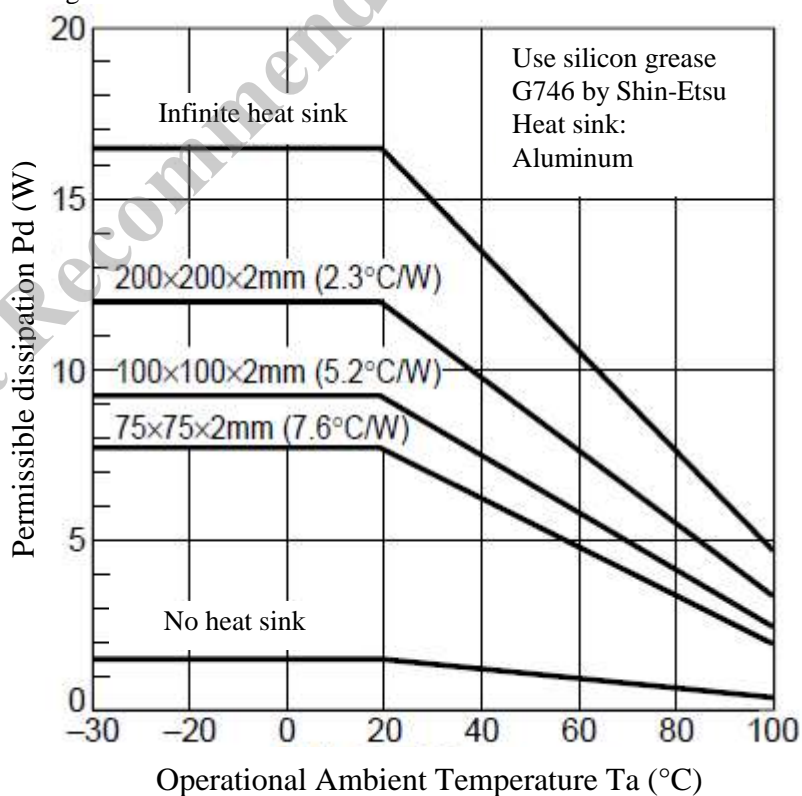
<u>Types</u>	<u>Suppliers</u>
G746	Shin-Etsu Chemical Co., Ltd.
SC102	Toray Silicone Co., Ltd.
YG6260	Momentive Performance Materials Inc.

Tightening torque of fixing screws

In order to keep the thermal resistance between the IC and the heat sink at low level without damaging the IC package, it is necessary to control the torque of fixing screws in a proper way. Even if silicon grease is coated, the thermal resistance θ_i increases if the tightening torque is not enough.

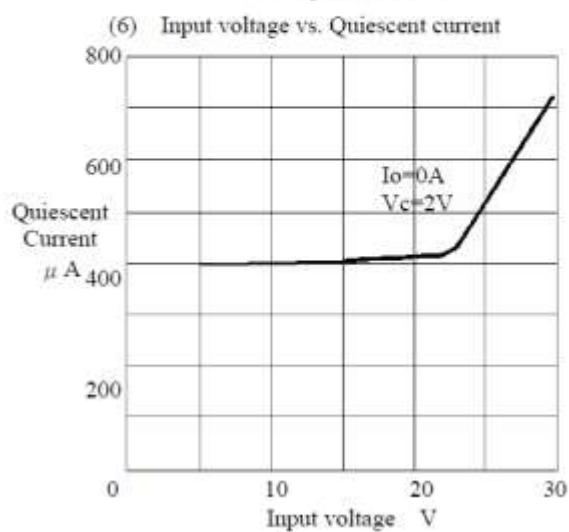
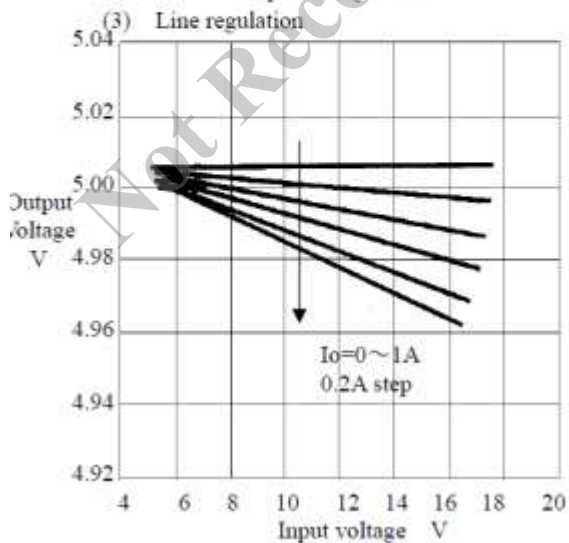
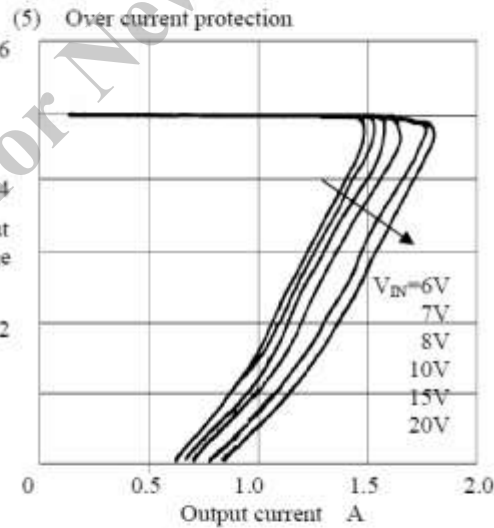
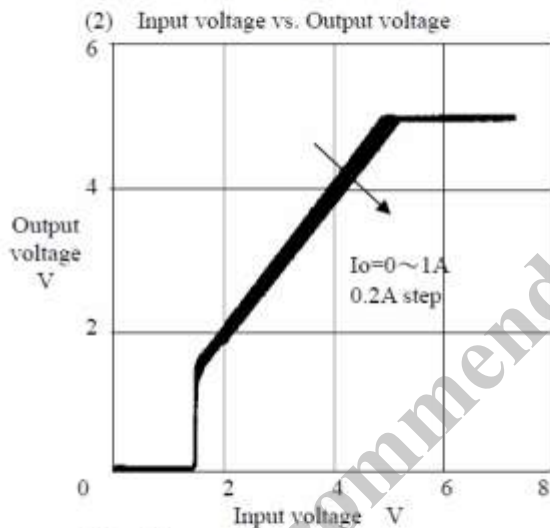
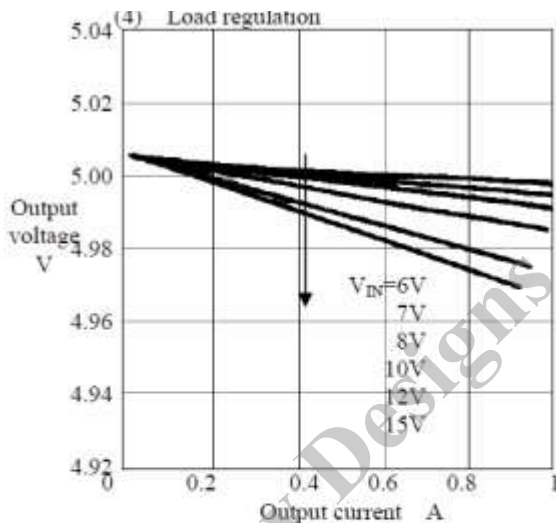
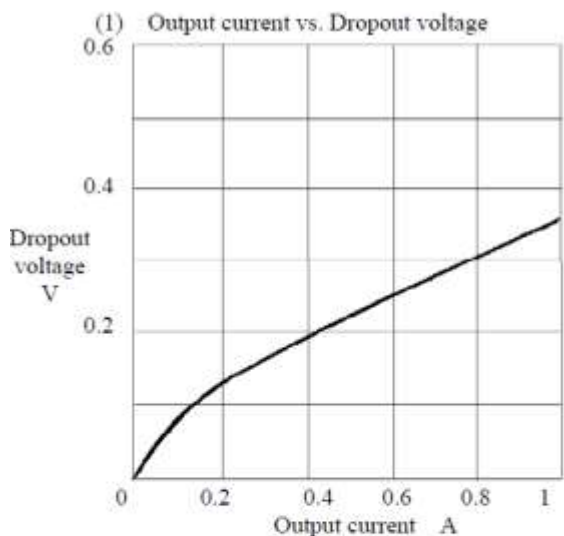
For the SI-3000KF, 58.8 - 68.6N cm (6.0 - 7.0kg cm) are recommended.

- SI-3000KF derating curve

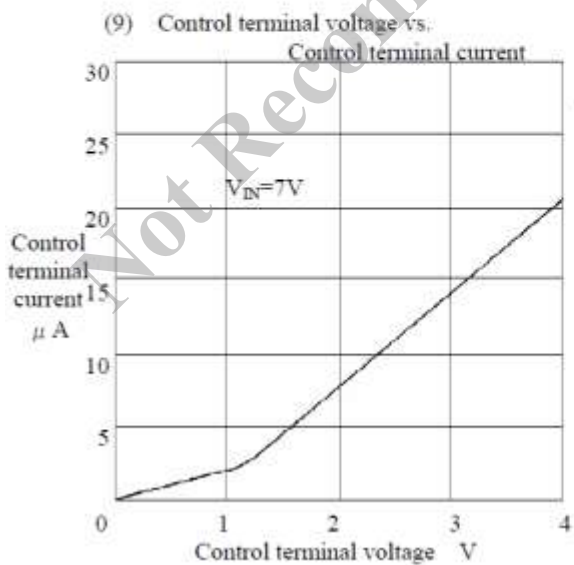
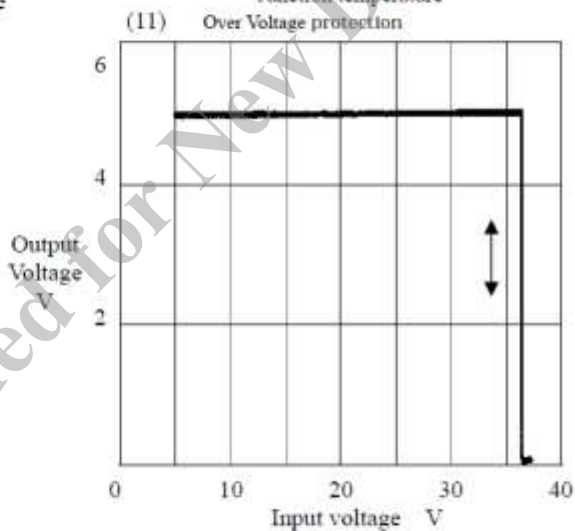
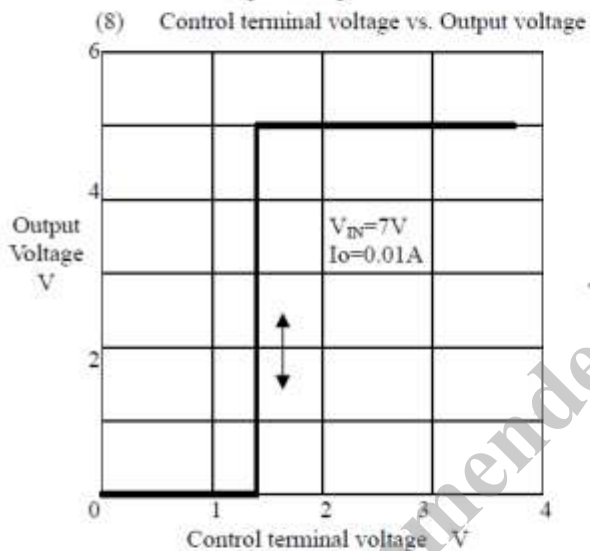
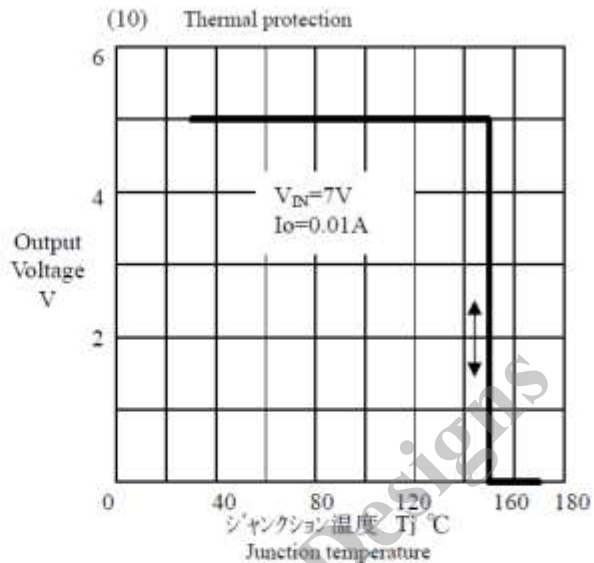
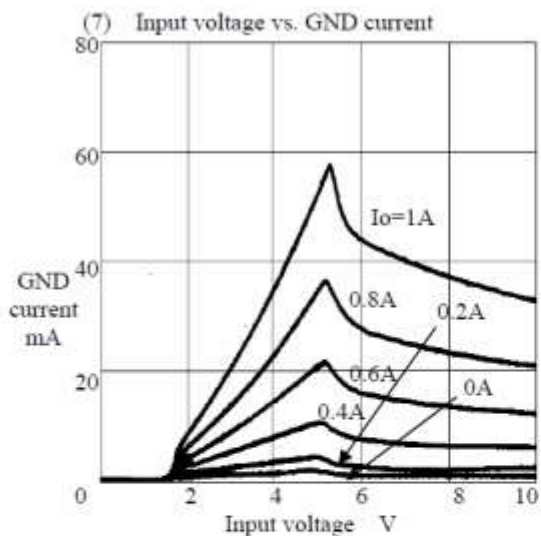


6. Typical Characteristics

(Ta = 25°C) *Set Vout = 5V (R2 = 10kΩ)



($T_a = 25^\circ\text{C}$) *Set $V_{out} = 5\text{V}$ ($R_2 = 10\text{k}\Omega$)



Notice

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