

$V_{POS} = -65\text{ V}$ ,  $R_{DS(ON)} = 120\text{ m}\Omega$   
**4 Series LED Bypass Switches**  
**SPF5047**

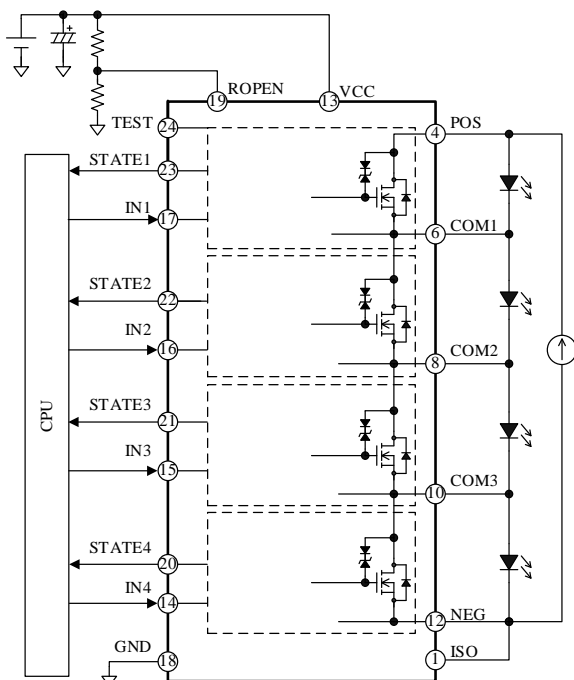
### Description

The SPF5047 is a bypass switch for the high-brightness matrix LEDs used in applications such as automotive headlights. The product includes four series switches to bypass individual LEDs, whose dimming systems are externally controlled by a CPU. Each LED has open and short detections that can send fault flags to the CPU. Even if one of the four LEDs becomes open, the open-LED is bypassed by its built-in power MOSFET so that other LEDs can continue to work normally. The SPF5047 is supplied in a compact, surface-mount type HSOP package, with an exposed pad for enhanced thermal dissipation.

### Features

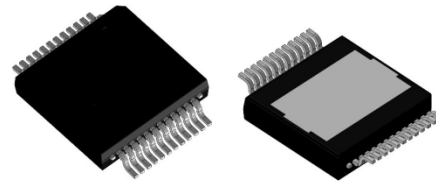
- 4 Series LED Bypass Switches
- Drives Up to 16 LEDs in Series (with 4 ICs; at  $V_F$  of an LED is approx. 3 V)
- Low Noise:  
No charge-current-induced noise occurs as each gate driver uses a negative power source which requires no charge-pump circuit.
- High Efficiency:  
Optimized trade-offs between switching loss and switching noise allow highly-accurate control in tr and tf.
- Fault Flag Reporting
- LED Open Detection
- LED Short Detection

### Typical Application



### Package

HSOP24



Not to scale

### Specifications

- Voltage across LED Capacitors in Series,  $V_{POS}$ :  
-65 V (max.)
- Bypass Switch Output Current,  $I_{OUT}$ : 2 A (max.)
- Bypass Switch On-Resistance (at 25 °C),  $R_{DS(ON)}$ :  
120 m $\Omega$  (typ.)

### Applications

- Automotive Headlight
- Daylight Running Lamp (DRL)
- High-Brightness Matrix LED System

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**1. Absolute Maximum Ratings**

Current polarities are defined as follows: current going into the IC (sinking) is positive current (+); and current coming out of the IC (sourcing) is negative current (-).

Unless specifically noted,  $T_A = 25\text{ }^\circ\text{C}$ , GND = ground.

Parameter	Symbol	Test Conditions	Rating	Unit	Remarks
VCC-NEG Pin Voltage	$V_{CN}$		70.5	V	
		$t < 10\ \mu\text{s}$	72	V	
NEG Pin Voltage	$V_{NEG}$		-65 to 0	V	
POS Pin Voltage	$V_{POS}$		-65 to 0	V	
VCC Pin Voltage	$V_{CC}$		0 to 5.5	V	
Input Pin Voltage	$V_{IN}$		0 to 5.5	V	
STATE Pin Voltage	$V_{ST}$		0 to 5.5	V	
STATE Pin Current	$I_{ST}$		$\pm 10$	mA	
ROPEN Pin Voltage	$V_{ROPEN}$		0 to 5.5	V	
MOSFET Drain-to-Source Voltage	$V_{DS}$		$V_{CLAMP}^*$	V	
MOSFET Active Clamp Time	$t_{CLAMP}$	1 channel; $I_{OUT} = 2\ \text{A}$	100	ms	
		4 channels at once; $I_{OUT} = 2\ \text{A}$	50	ms	
Output Current	$I_{OUT}$		2	A	
Junction Temperature	$T_j$		150	$^\circ\text{C}$	
Operating Ambient Temperature	$T_{OPR}$		-40 to 125	$^\circ\text{C}$	
Power Dissipation	$P_D$		6.2	W	
Storage Temperature	$T_{stg}$		-40 to 150	$^\circ\text{C}$	

\* Limited by the active clamp voltage,  $V_{CLAMP}$ .

**2. Electrostatic Discharge (ESD)**

ESD immunity test conditions:  $T_A = 25\text{ }^\circ\text{C}$ , ISO = ground.

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Human Body Model (HBM)	$V_{ESD(HBM)}$	$C = 100\ \text{pF}$ , $R = 1.5\ \text{k}\Omega^*$	-2000	—	2000	V
Machine Model (MM)	$V_{ESD(MM)}$	$C = 200\ \text{pF}$ , $R = 0\ \Omega$ ; all pins	-200	—	200	V

\* Based on JEITA EIAJ ED-4701/304.

### 3. Recommended Operating Conditions

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit	Remarks
Logic Supply Voltage	$V_{CC}$		4.9	—	5.1	V	VCC pin
PWM Drive Frequency	$f_{PWM}$		100	—	500	Hz	IN1 to IN4 pins

### 4. Electrical Characteristics

Current polarities are defined as follows: current going into the IC (sinking) is positive current (+); and current coming out of the IC (sourcing) is negative current (-). Unless specifically noted, the characteristic values listed in this section are values per channel, and GND is the ground. Unless otherwise specified, the conditions are as follows:  $T_j = -40\text{ }^{\circ}\text{C}$  to  $125\text{ }^{\circ}\text{C}$ ; <sup>(1)</sup> the VCC pin voltage,  $V_{CC} = 5.0\text{ V}$ ; the NEG pin voltage,  $V_{NEG} = -30\text{ V}$ ; with the NEG and ISO pins being shorted. The voltage across the INx pin, which is placed in the channel not to be measured, should be fixed at 5 V.

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Operation Start Voltage	$V_{CC(ON)}$		3.0	3.5	4.0	V
Operation Stop Voltage	$V_{CC(OFF)}$		2.4	2.7	3.0	V
Circuit Current 1	$I_{CC1}$	$V_{IN1} = V_{IN2} = V_{IN3} = V_{IN4} = 0\text{ V}$	—	—	11	mA
Circuit Current 2	$I_{CC2}$	$V_{IN1} = V_{IN2} = V_{IN3} = V_{IN4} = 5\text{ V}$ ; drain-to-source voltage per MOSFET = 4 V	—	—	11.5	mA
IN Pin Input Threshold Voltage	$V_{THH}$		3.5	—	—	V
	$V_{THL}$		—	—	1.5	V
IN Pin Input Current	$I_{IN}$		—	50	100	$\mu\text{A}$
Output On-Resistance	$R_{DS(ON)}$	$T_j = 25\text{ }^{\circ}\text{C}$ , $I_{OUT} = 1\text{ A}$	—	120	150	m $\Omega$
		$T_j = 125\text{ }^{\circ}\text{C}$ , $I_{OUT} = 1\text{ A}$	—	—	250	m $\Omega$
Active Clamp Voltage	$V_{CLAMP}$	$I_{OUT} = 1\text{ A}$ ; INx in the non-measuring channel = 0 V fixed	12	15	18	V
Short Detection Voltage <sup>(2)</sup>	$V_{SH}$	$V_{ROPEN} = 2\text{ V}$	1.25	1.5	1.75	V
Setting Range for Open Detection Threshold Voltage <sup>(2)</sup>	$V_{OPR}$		1.75	—	11	V
Open Detection Voltage <sup>(2)</sup>	$V_{OP2}$	$V_{ROPEN} = 0.5\text{ V}$	1.84	2.00	2.16	V
	$V_{OP4}$	$V_{ROPEN} = 1\text{ V}$	3.68	4.00	4.32	V
	$V_{OP8}$	$V_{ROPEN} = 2\text{ V}$	7.36	8.00	8.64	V
	$V_{OP10}$	$V_{ROPEN} = 2.5\text{ V}$	9.2	10.0	10.8	V
ROPEN Pin Input Current	$I_{ROP}$	$V_{ROPEN} = 3\text{ V}$	—	1.2	4	$\mu\text{A}$
STATE Pin Output Voltage	$V_{STH}$	$I_{STATE} = -5\text{ mA}$	$V_{CC} - 0.5$	—	$V_{CC}$	V
	$V_{STL}$	$I_{STATE} = 5\text{ mA}$	0	—	0.5	V
Minimum Open/Short Detection Time	$t_{DET}$		50	—	—	$\mu\text{s}$

<sup>(1)</sup> " $T_j = -40\text{ }^{\circ}\text{C}$ " is the parameter ensured by design only. For shipping inspection, all electrical characteristics values should be evaluated at  $T_j = 25\text{ }^{\circ}\text{C}$  or  $125\text{ }^{\circ}\text{C}$ .

<sup>(2)</sup> Refers to drain-to-source voltage across each built-in MOSFET.

# SPF5047

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Internal Oscillator Clock Cycle	$t_{CLK}$		3.04	4	4.96	$\mu s$
Short Detection Filter Time	$t_{SH}$	$V_{ROPEN} = 2 V$	17.1	22.5	27.9	ms
Open Detection Filter Time	$t_{OP}$	$V_{ROPEN} = 2 V$	17.1	22.5	27.9	ms
Short Detection Recovery Delay Time	$t_{SHB}$	$V_{ROPEN} = 2 V$	—	—	50	$\mu s$
Open Detection Delay Time	$t_{OPR}$	$V_{ROPEN} = 2 V$	—	—	50	$\mu s$
Open Recovery Filter Time	$t_{OPB}$	$V_{ROPEN} = 2 V$	12.4	16.4	20.4	ms
Filter Time Difference	$t_{OP-OPB}$	$t_{OP-OPB} = t_{OP} - t_{OPB}$	4.7	6.1	7.5	ms
Output Propagation Delay	$t_{ON}$	(3)	—	40	80	$\mu s$
	$t_{OFF}$		—	40	80	$\mu s$
Output Rise Time	$t_r$		30	70	110	$\mu s$
Output Fall Time	$t_f$		30	70	110	$\mu s$
Output Propagation Delay Difference	$t_{DIFF}$		-30	—	30	$\mu s$
Junction-to-Case Thermal Resistance	$\theta_{jC}$			—	—	4.9

(3) Drain-to-source voltage per MOSFET = -30 V, with a 4  $\Omega$  source resistor being connected. Short the NEG pin to the source pin in the channel to be measured. For switching characteristics definitions, see Figure 4-1 below.

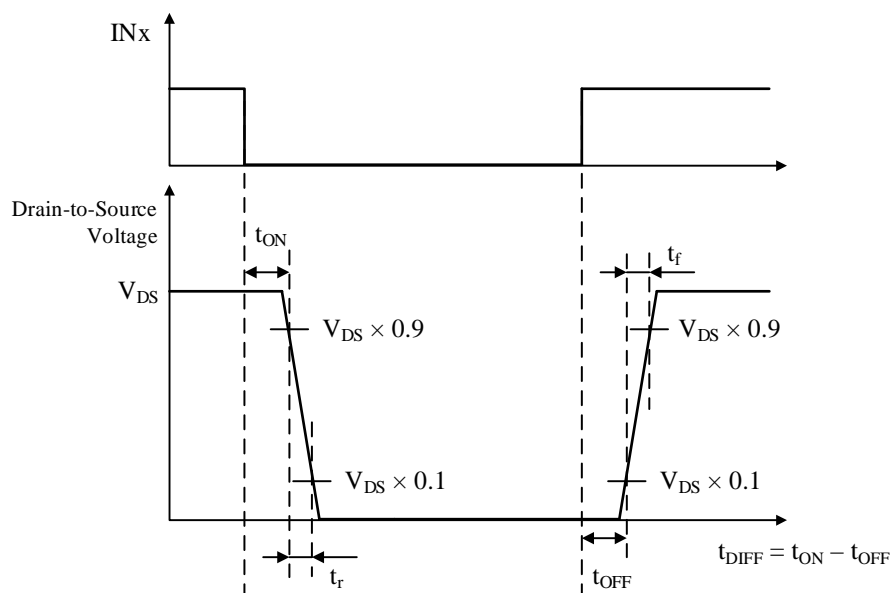
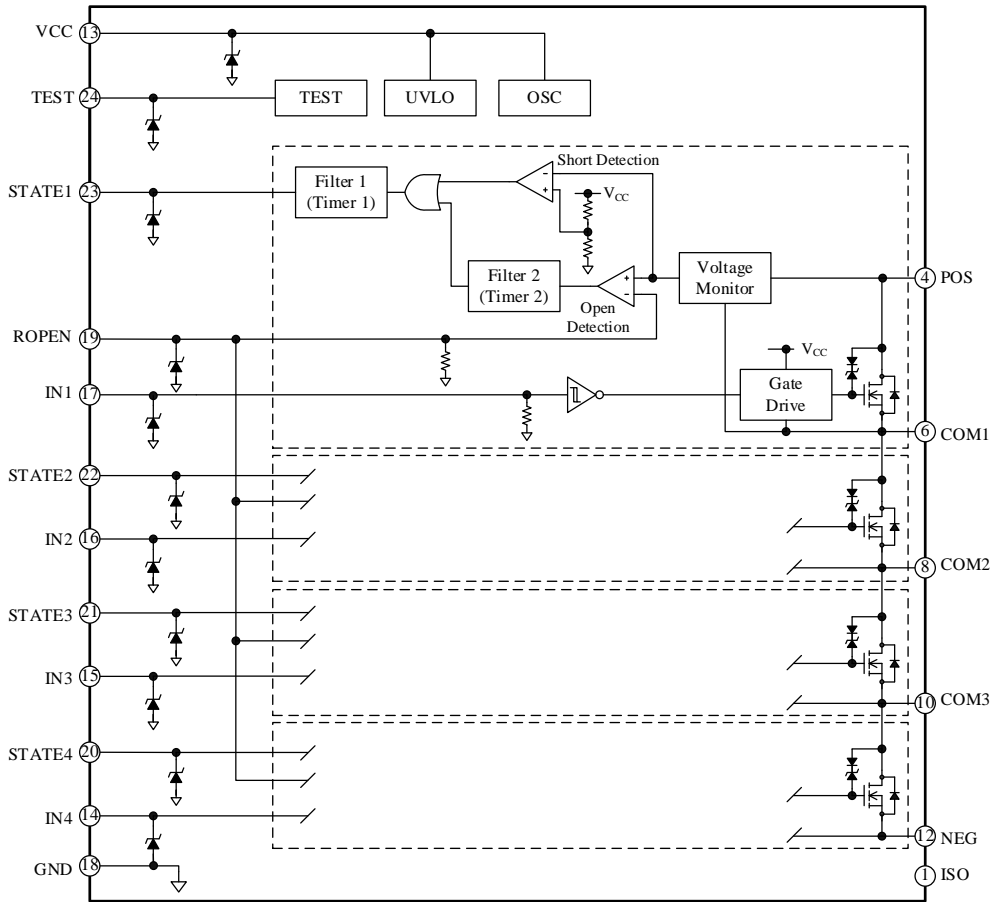
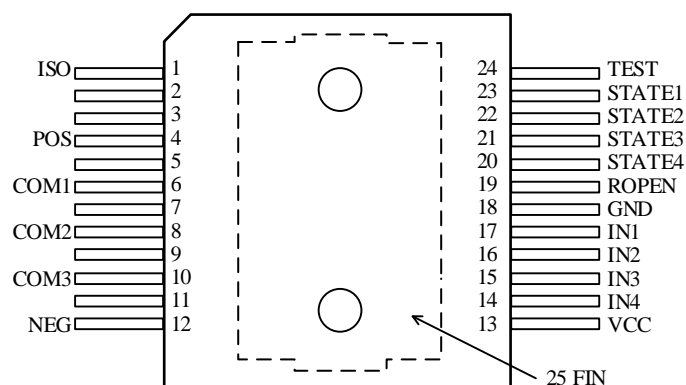


Figure 4-1. Switching Characteristics Definitions

5. Block Diagram



## 6. Pin Configuration and Definitions



Pin Number	Pin Name	Function
1	ISO	MIC reference; internally shorted to NEG
2	N.C.	No connection
3	N.C.	No connection
4	POS	Drain of Channel 1
5	N.C.	No connection
6	COM1	Source of Channel 1, drain of Channel 2
7	N.C.	No connection
8	COM2	Source of Channel 2, drain of Channel 3
9	N.C.	No connection
10	COM3	Source of Channel 3, drain of Channel 4
11	N.C.	No connection
12	NEG	Source of Channel 4; internally shorted to ISO
13	VCC	5 V power supply
14	IN4	Input for Channel 4
15	IN3	Input for Channel 3
16	IN2	Input for Channel 2
17	IN1	Input for Channel 1
18	GND	Ground
19	ROOPEN	Threshold voltage setting for LED open detection; commonly applied to all channels (1 to 4)
20	STATE4	Open/short detection for Channel 4
21	STATE3	Open/short detection for Channel 3
22	STATE2	Open/short detection for Channel 2
23	STATE1	Open/short detection for Channel 1
24	TEST	Product inspection; normally left unused; must be shorted to GND when used
25	FIN	Backside heatsink; must be shorted to NEG and ISO when used

### 7. Typical Applications

The NEG pin, the ISO pin, and the backside heatsink must be shorted on a PCB. Connect these pins to the lowest potential (negative potential) among other pins.

The TEST pin must be shorted to the GND pin on a PCB.

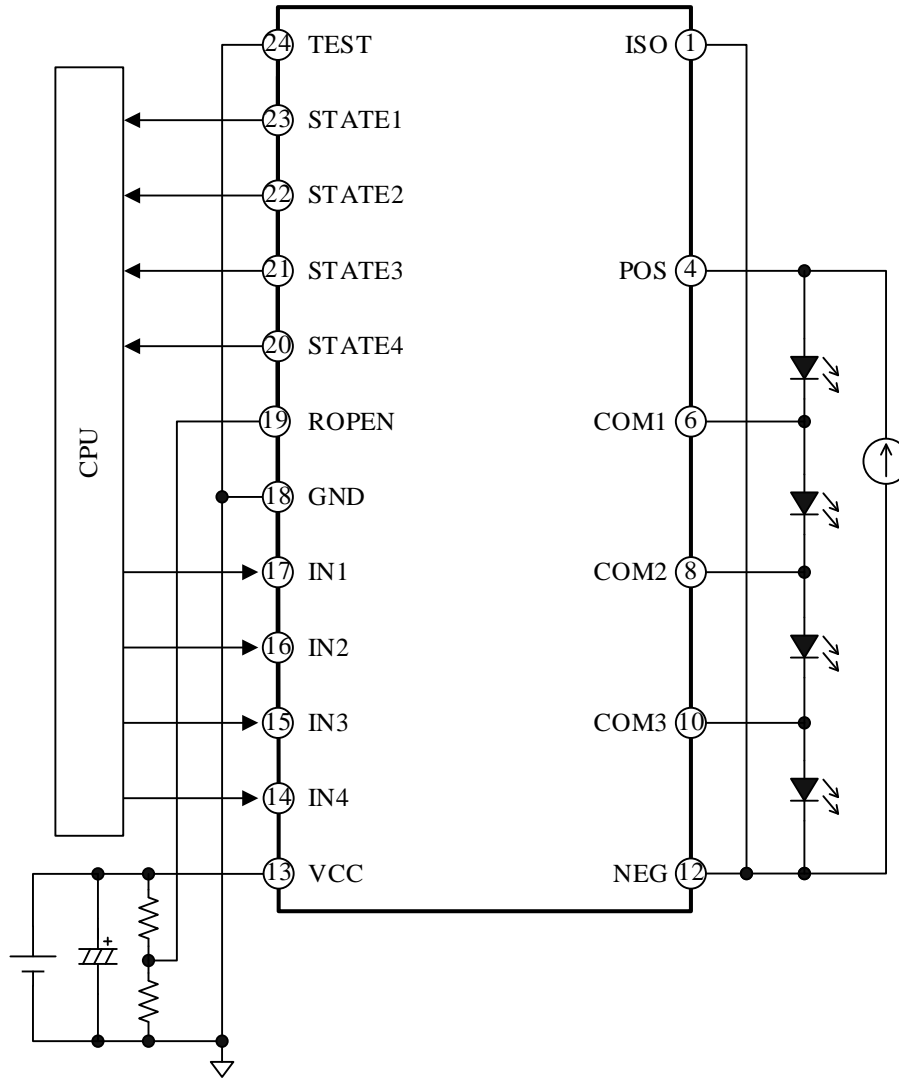


Figure 7-1. Typical Application (4 channels)



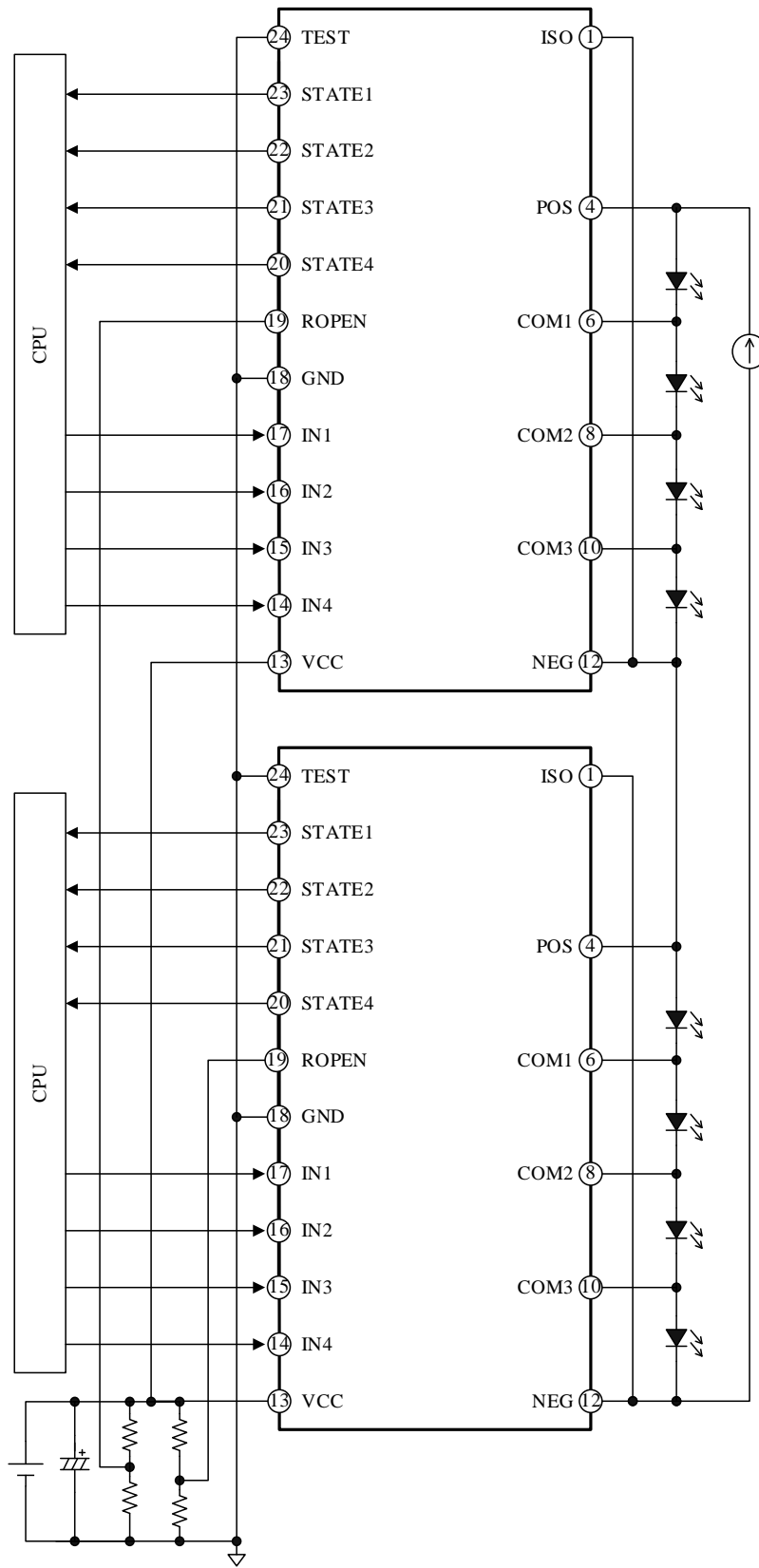
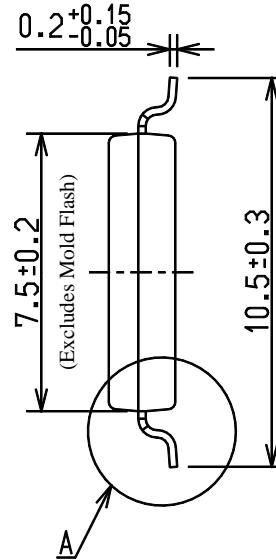
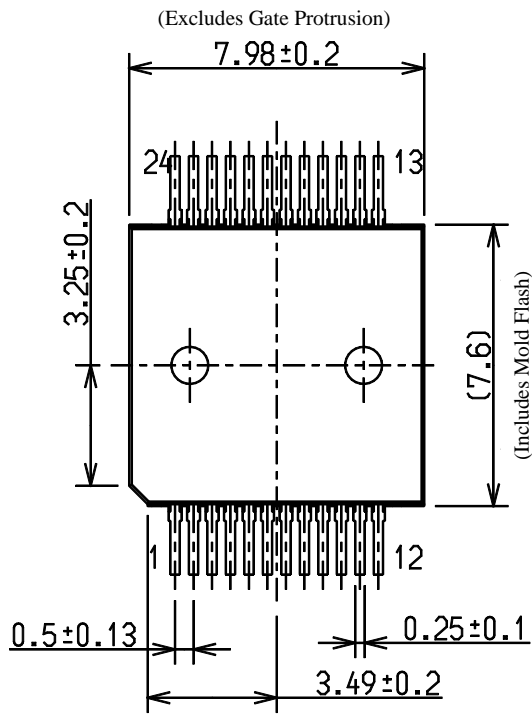


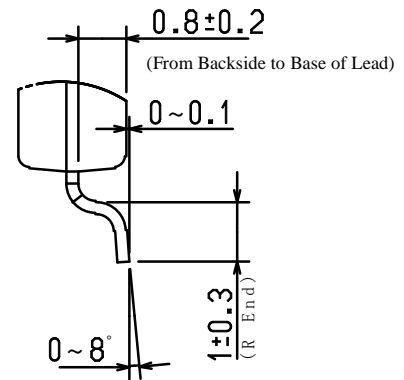
Figure 7-2. Typical Application (8 channels)

8. Physical Dimensions

- HSOP24

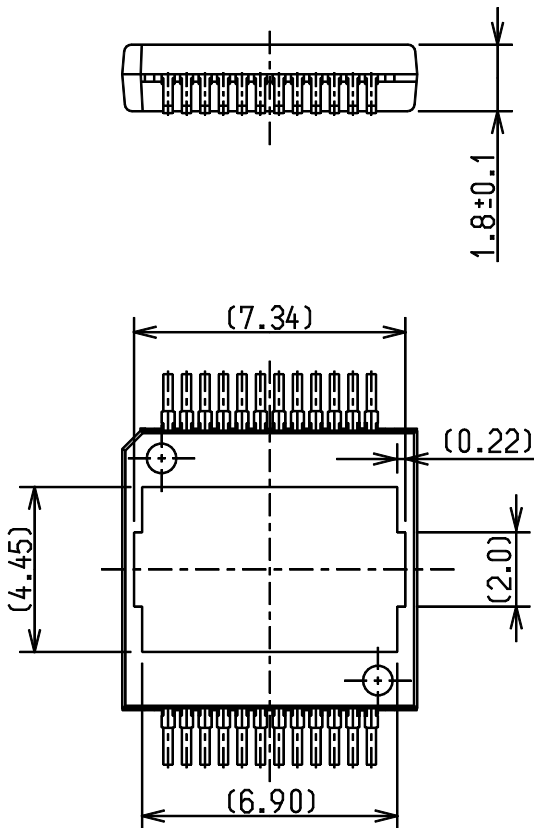


Enlarged View of A

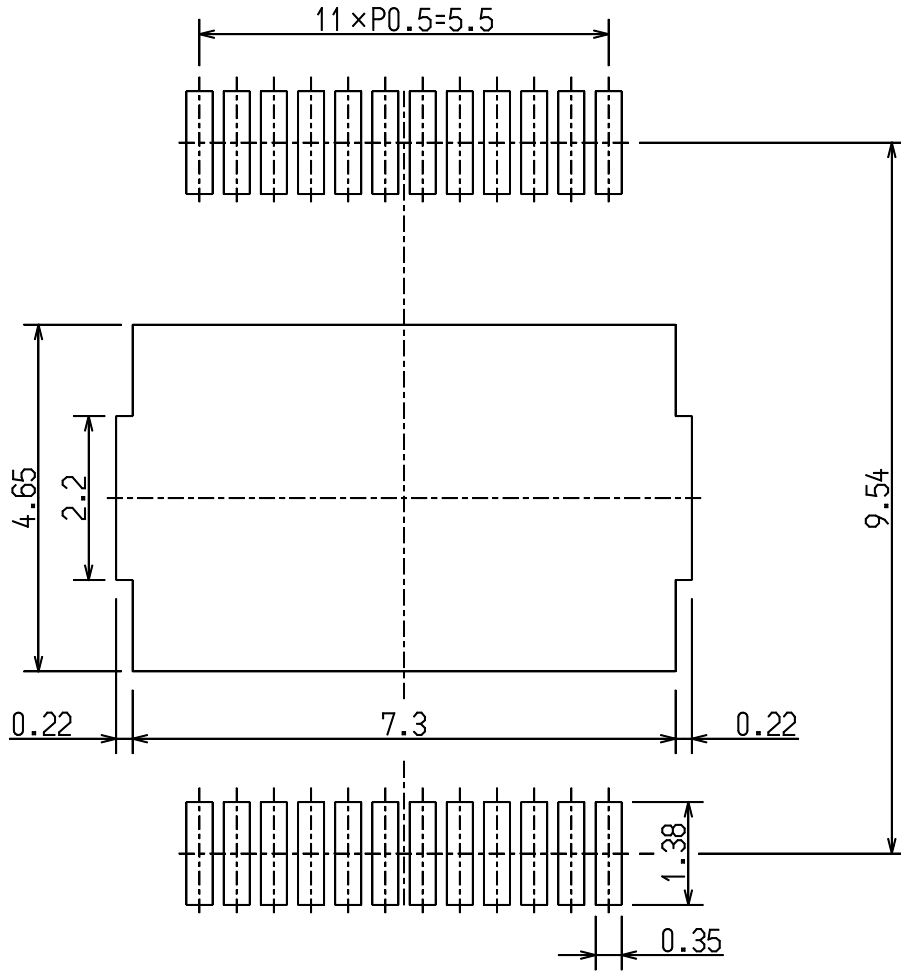


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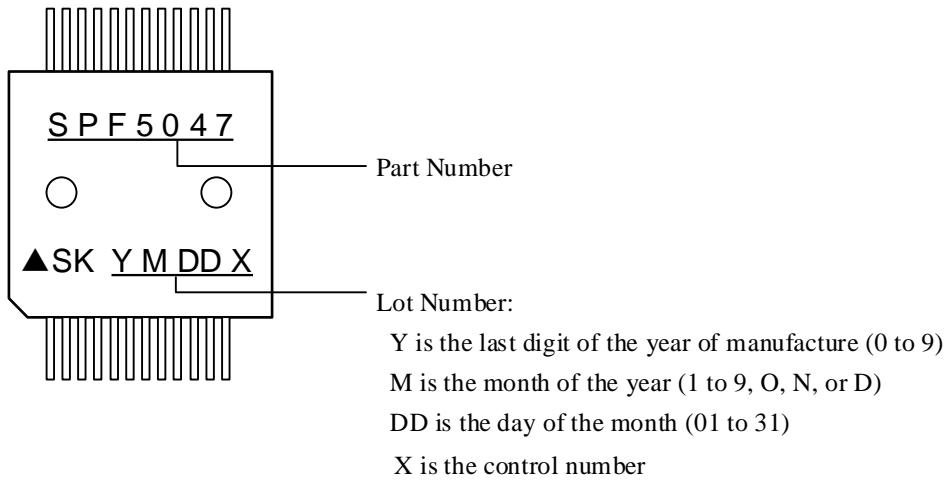
- Dimensions in millimeters
- Pin treatment: Pb-free (RoHS compliant)
- Reflow



9. Recommended Land Pattern



10. Marking Diagram



## 11. Operational Description

The characteristic values listed in this section are typical values, unless they are specified as minimum or maximum. Current polarities are defined as follows: current going into the IC (sinking) is positive current (+); and current coming out of the IC (sourcing) is negative current (-).

### 11.1. Overview

The SPF5047 is designed to provide dynamic dimming control for each of the four LEDs connected in series. The built-in MOSFETs connected in parallel to the LEDs alter the currents through these LEDs individually. With this array strategy, the LEDs used in the SPF5047 can be quickly turned off and be partially dimmed even they are series-connected.

Moreover, each LED has open and short detections that can emit fault flags.

### 11.2. Turning On and Off the IC

The IC includes the UVLO (Undervoltage Lockout) circuit. When  $V_{CC(ON)} \geq 3.5$  V, the control circuit in the IC is activated.

The IC then drives the LEDs and sends PWM signals to all the INx pins (IN1 to IN4).

When  $V_{CC(OFF)} \leq 2.7$  V, the UVLO circuit starts its operation to deactivate the control circuit.

### 11.3. Switching Operation

A driving signal on the INx pin actuates the corresponding output MOSFET. (“INx” represents any one of the outputs.) The output MOSFET then starts operating synchronously with the input signal on the INx pin. In a state where the INx pin is logic low (or “L”) and the MOSFET is on, an individual LED turns off. The INx pin is also used to control LED brightness. A PWM signal on the INx pin can enable PWM dimming, which changes the light intensity of an LED with a predetermined duty cycle. The frequency of PWM signals should range from 100 Hz to 500 Hz.

The IC allows highly-accurate control in switching characteristics,  $t_r$  and  $t_f$ , through optimized trade-offs between switching loss and switching noise. Consequently, the IC achieves high-efficient, low-noise switching operations.

The control circuit is driven by a positive power source, whereas the MOSFET driver is actuated by a negative power source in which the voltage between the POS and NEG pins is lower than the ground potential. In this method, the MOSFET driver does not need any charge-pump circuit; therefore, charge-current-induced noise can be suppressed.

## 11.4. LED Open/Short Detection

Figure 11-1 shows the internal block diagram of Channel 1 (CH1).

The IC monitors the output MOSFET drain-to-source voltage,  $V_{DS}$ , to determine if an LED open or short event has occurred.

First,  $V_{DS}$  is measured and then multiplied by 0.25, with an internal voltage monitor. Then, the 0.25 times voltage value is input to open and short detection comparators, respectively. Each of the open and short detection comparators has a filter with the minimum open/short detection time,  $t_{DET}$ , of 50  $\mu$ s (min.). Therefore, a signal shorter than  $t_{DET}$  cannot be detected as a fault flag.

Figure 11-2 describes the relationship between input voltage across the open/short detection comparator and drain-to-source voltage.

When the IC detects any malfunction, the STATEx pin is set to logic high (or “H”) and transmits a fault flag to the microcontroller. (“STATEx” represents any one of the outputs.) The IC incorporates filters to remove noise signals so that the microcontroller can properly respond to a low-to-high transition on the STATEx pin input signal. A clock cycle generated by the built-in oscillator (OSC) determines a filter time.

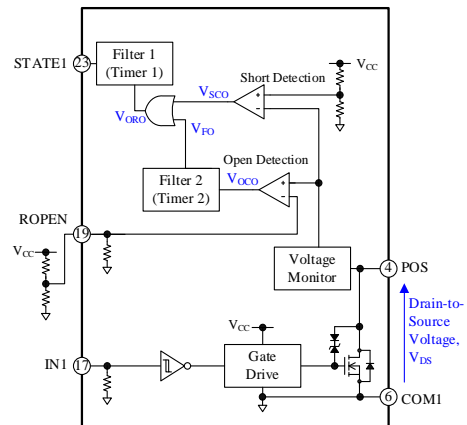


Figure 11-1. Internal Block Diagram (CH1)

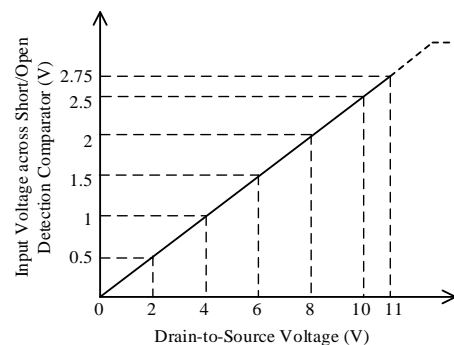


Figure 11-2. Input Voltage across Open/Short Detection Comparator vs. Drain-to-Source Voltage

### 11.4.1. LED Short Detection

Figure 11-3 shows the operating waveforms during an LED short event. And Figure 11-4 provides the enlarged view of A, illustrated in Figure 11-3.

If one or more LEDs are shorted to ground, the MOSFET drain-to-source voltage,  $V_{DS}$ , attempts to decrease.

The short detection voltage,  $V_{SH}$ , is used for LED status diagnostics and is set to 1.5 V. When a short condition (where  $V_{DS} \leq V_{SH}$ ) persists for a period greater than the short detection filter time ( $t_{SH}$ ) of 22.5 ms, the STATEx pin in the shorted-LED channel becomes logic high (or “H”).

When  $V_{DS}$  exceeds  $V_{SH}$  under the short condition, the short detection comparator waits to invert its output until  $t_{SHB}$ , the open detection recovery time of 50  $\mu$ s (max.), has elapsed. Following the output inversion of the short detection comparator, the STATEx pin is put into a logic low state (or “L”) after 12 cycles of a clock pulse. Then, the IC resumes its normal operation.

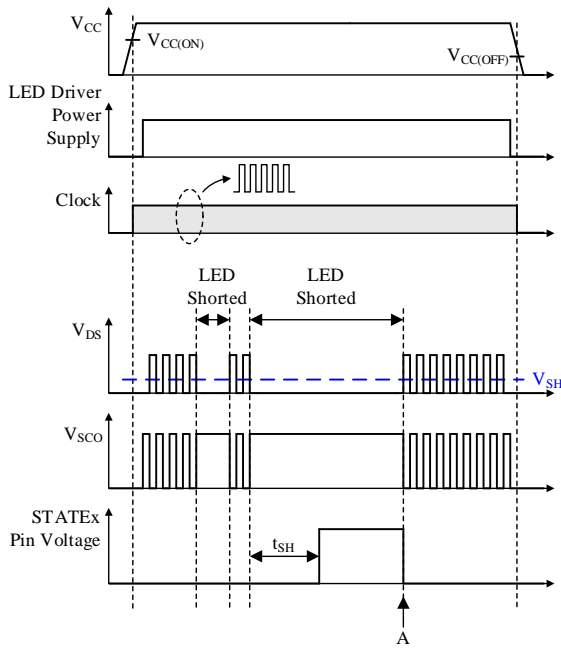


Figure 11-3. Waveforms during Short Detection

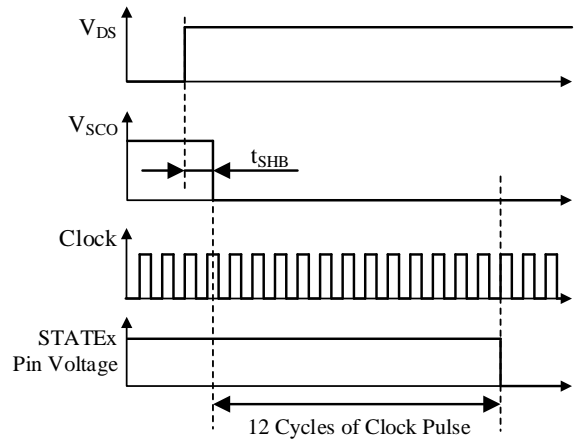


Figure 11-4. Waveforms during Short Detection (Enlarged View of A)

### 11.4.2. LED Open Detection

Figure 11-5 shows the operating waveforms during an LED open event. And Figure 11-6 provides the enlarged view of B, illustrated in Figure 11-5.

If one or more LEDs are open, the MOSFET drain-to-source voltage,  $V_{DS}$ , attempts to increase.

The open detection voltage,  $V_{OP}$ , is used to detect open-LEDs. When an open condition (where  $V_{DS} \geq V_{OP}$ ) persists for a period greater than the open detection filter time ( $t_{OP}$ ) of 22.5 ms, the STATEx pin in the open-LED channel becomes logic high (or “H”).

Even after the open detection comparator output,  $V_{OCO}$ , is held logic high, the microcontroller does not respond to an “H” signal whose duration is shorter than 3 cycles of a clock pulse (see Figure 11-5). This is to prevent false detections. When  $V_{DS}$  falls below  $V_{OP}$ , the voltage on the open detection comparator,  $V_{OCO}$ , will be inverted after  $t_{OPB}$ , the open recovery filter time of 16.4 ms. Following the output inversion of the open detection comparator, the short detection comparator is inverted after  $t_{SHB}$  of 50  $\mu$ s (max.). Then, the STATEx pin is put into a logic low state (or “L”) after 12 cycles of a clock pulse. This high-to-low transition mechanism is identical to that of the LED short detection, as shown in Figure 11-4. Then, the IC resumes its normal operation.

Table 11-1 indicates how the  $V_{OP}$  values vary according to the ROPEN pin voltages. Hence, note that the ROPEN pin voltages must be determined according to LED currents.

Table 11-1. Open Detection Voltage vs. ROPEN Pin Voltage

Symbol	ROPEN Pin Voltage	Open Detection Voltage, $V_{OP}$		
		Min.	Typ.	Max.
$V_{OP2}$	0.5 V	1.84 V	2.00 V	2.16 V
$V_{OP4}$	1 V	3.68 V	4.00 V	4.32 V
$V_{OP8}$	2 V	7.36 V	8.00 V	8.64 V
$V_{OP10}$	2.5 V	9.2 V	10.0 V	10.8 V

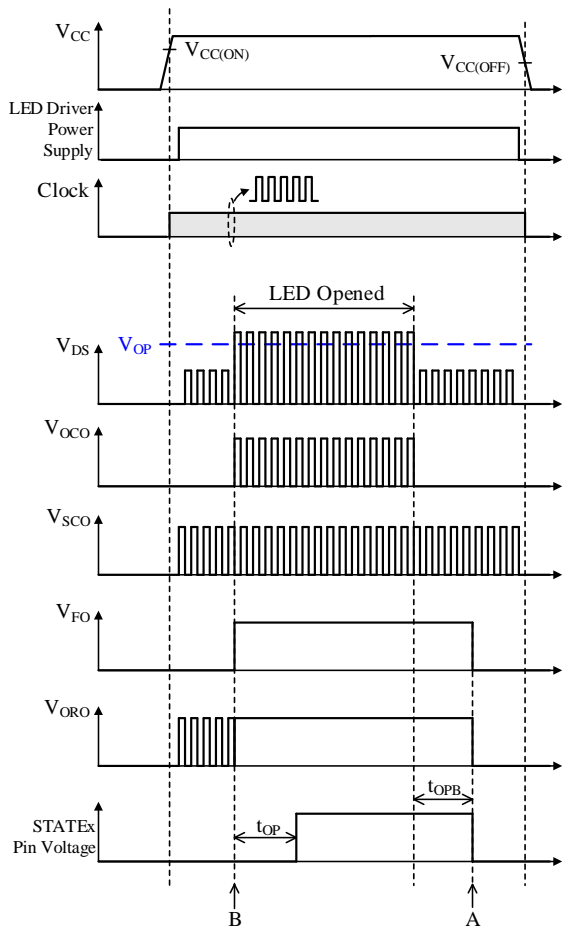


Figure 11-5. Waveforms during Open Detection

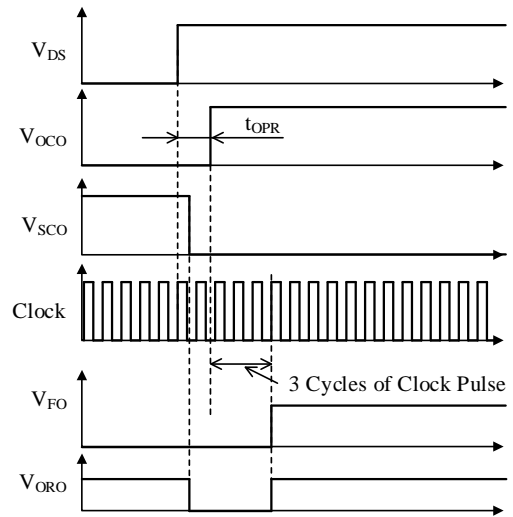


Figure 11-6. Waveforms during Open Detection (Enlarged View of B)

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